



Dept. of Electrical, Computer and Biomedical Engineering

Data acquisition from a photodiode

Photodiodes

Photodiodes are semiconductor devices with PN or PIN structure typically used as radiant power transducers

- The energy transferred by the electromagnetic radiation, absorbed in the depletion or in the intrinsic region is responsible for the generation of electron/hole pairs, eventually contributing to the formation of a current
- The voltage/current characteristic of a photodiode is therefore the same as in a diode, with the addition of a photo-generated current term I_{ph}

$$\mathbf{I}_{\mathrm{D}} = \mathbf{I}_{\mathrm{0}} \left(\mathbf{e}^{\mathsf{V}_{\mathrm{D}}/\mathsf{V}_{\mathrm{T}}} - \mathbf{1} \right) - \mathbf{I}_{\mathrm{ph}} \qquad \qquad \mathbf{I}_{\mathrm{D}} \left| \mathbf{\Psi}^{+} \right|$$

where I_0 is the diode leakage current, V_D is the voltage across the device and V_T is the thermal voltage. Note that, in reverse bias operating conditions (V_D <0), the first term in the expression reduces to I_0 , while for V_D =0, I_D =- I_{ph} .

 V_{D}

Photodiodes

The photo-generated current I_{ph} is proportional to the incident radiant power, i.e., to the flux of photons hitting the device:

$$I_{ph} = S \cdot P = \frac{\eta e}{h\nu} P$$
, $\frac{P}{h\nu} = \#$ fotoni al sec.

- where S is the spectral (or radiometric) sensitivity, η is the quantum efficiency, **e** is the elementary charge (1.602 10⁻¹⁹ C), **P** is the power of the incident electromagnetic wave, **h** is the Plank's constant (6.625 10⁻³⁴ J·s) and ν is the electromagnetic wave frequency
- Other characteristic parameters of a photodiode are the linearity, the dark current, the junction capacitance, the breakdown voltage and the response time



Main uses of thermistors

Application field	Use or device				
Cameras	Light intensity measurement, automatic control of the shutter, auto- focus, flash unit control				
Medical instrumentation	TAC scanner, X-ray detection, biological analysis (e.g., blood), oximetry				
Safety devices	Smoke and flame detectors, X-ray systems for airplane inspection, intrusion detectors				
Automotive	Headlight dimmer, sun light detector (for air conditioning)				
Communications	Opto-electronic converters, remote optical control				
Industry	Bar code readers, encoders, position detectors, toner density measurement in printers				

Purpose of the experiment

- Implement a system for data acquisition from a radiant power transducer, in particular from a photodiode. The system should include
 - a conditioning circuit for the signal coming from the photodiode
 - a virtual instrument implemented in the LabView programming environment serving as an interface between the acquisition system and the user
- The virtual instrument should take care of acquiring the room radiant power and representing the time evolution of the measured voltage and the instantaneous value of the radiant power

Operating modes for the photodiode

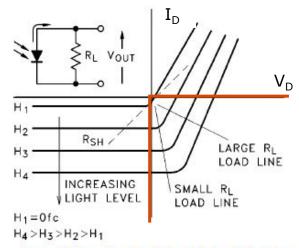
Photovoltaic mode: the photodiode is operated with no bias voltage applied and can supply electrical power (in the passive sign convention, $V_D I_D < 0$, with $I_D \le 0$ and $V_D > 0$); in particular, for $I_D = 0$, the photodiode behaves like a voltage source

$$V_{\text{D}} = V_{\text{T}} In \! \left(\frac{I_{\text{ph}}}{I_0} + 1 \right) \label{eq:V_D}$$

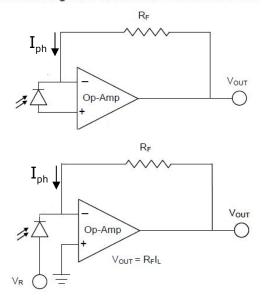


Photoconductive mode: the photodiode is operated in reverse or zero volt bias conditions, $V_{D} \leq 0$, and behaves like a current source; in particular, if the potential difference across the device is close to zero

$$I^{}_{\text{D}} = -I^{}_{\text{ph}}$$

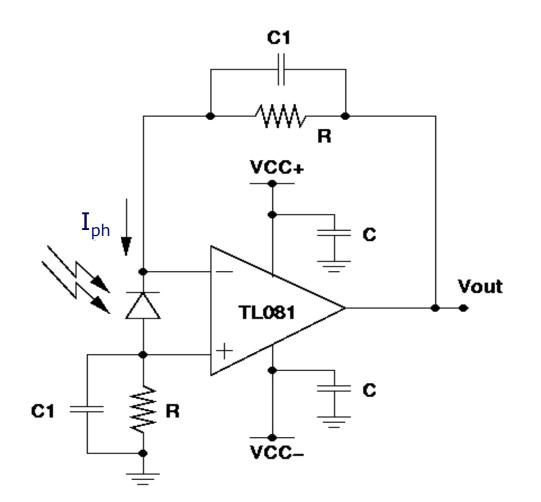


Current/Voltage Characteristics - Photovoltaic Mode





Conditioning circuit



- R=1 M Ω
- C=100 nF
- C1=15 nF
- VCC+=+15 V
- VCC⁻=-15 V
- PD: VTB8440B

Conditioning circuit

The proposed scheme, in principle, makes it possible to cancel the dark current contribution (current flowing through the diode when V_D=0)

$$V_{out} = 2 \cdot R \cdot I_{ph} = 2 \cdot S \cdot R \cdot P_{\text{incident light power spectral sensitivity}}$$

Another advantage of the proposed conditioning circuit lies in the reduction of the effects of the input bias currents on the output of the operational amplifier

$$V_{out}|_{I^+,I^-} = -I^+R + I^-R = -R\Delta I$$

where $\Delta I = I^+ - I^-$ is the offset of the input bias currents of the operational amplifier

TLO81 JFET input OpAmp

National Semiconductor

December 1995

TL081 Wide Bandwidth JFET Input Operational Amplifier

General Description

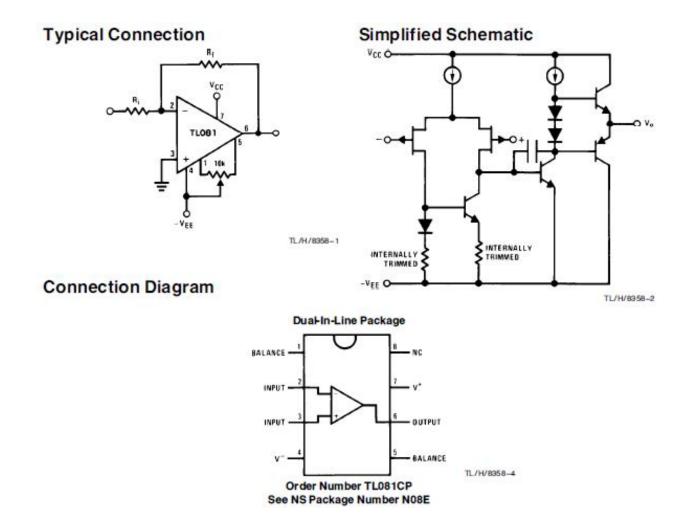
The TL081 is a low cost high speed JFET input operational amplifier with an internally trimmed input offset voltage (BI-FET IITM technology). The device requires a low supply current and yet maintains a large gain bandwidth product and a fast slew rate. In addition, well matched high voltage JFET input devices provide very low input bias and offset currents. The TL081 is pin compatible with the standard LM741 and uses the same offset voltage adjustment circuitry. This feature allows designers to immediately upgrade the overall performance of existing LM741 designs.

The TL081 may be used in applications such as high speed integrators, fast D/A converters, sample-and-hold circuits and many other circuits requiring low input offset voltage, low input bias current, high input impedance, high slew rate and wide bandwidth. The devices has low noise and offset voltage drift, but for applications where these requirements are critical, the LF356 is recommended. If maximum supply current is important, however, the TL081C is the better choice.

Features

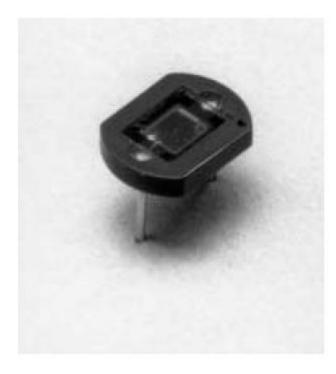
Internally trimmed offset voltage	15 mV
Low input bias current	50 pA
Low input noise voltage	25 nV/√Hz
Low input noise current	0.01 pA/ _v Hz
Wide gain bandwidth	4 MHz
 High slew rate 	13 V/µs
Low supply current	1.8 mA
 High input impedance 	10 ¹² Ω
• Low total harmonic distortion $A_V = 10$, $R_L = 10k$, $V_O = 20$ Vp-p, BW = 20 Hz-20 kHz	<0.02%
Low 1/f noise corner	50 Hz
Fast settling time to 0.01 %	2 µs

TLO81 JFET input OpAmp

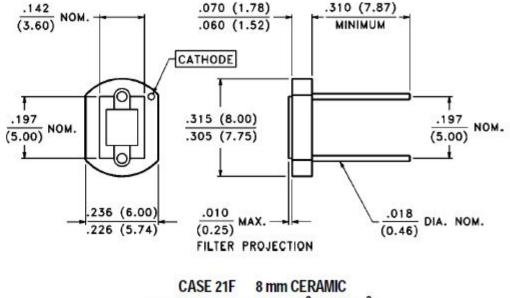




VTB8440B photodiode



PACKAGE DIMENSIONS inch (mm)



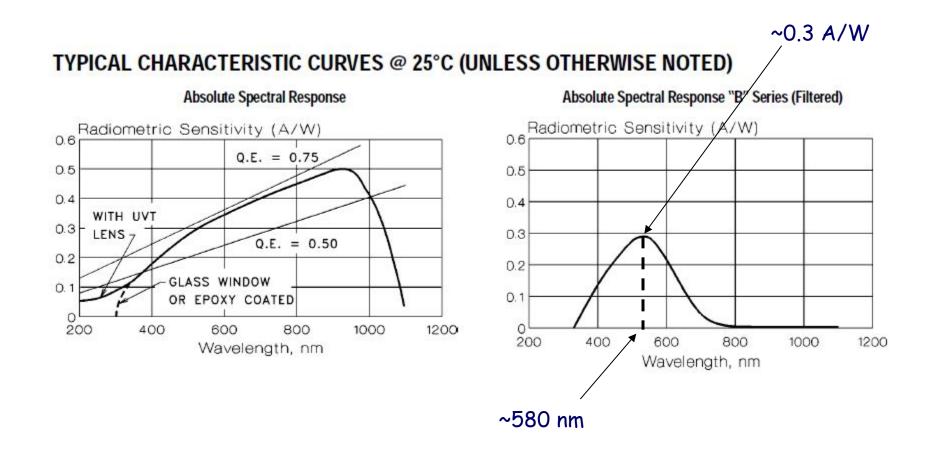
CHIP ACTIVE AREA: .008 in² (5.16 mm²)

VTB8440B photodiode

ELECTRO-OPTICAL CHARACTERISTICS @ 25°C (See also VTB curves, pages 21-22)

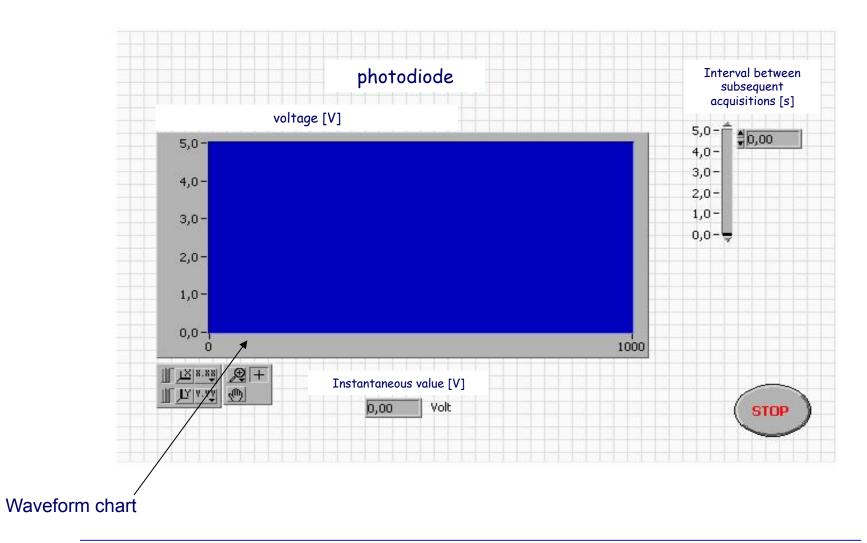
SYMBOL	CHARACTERISTIC	TEST CONDITIONS	VTB8440B			VTB8441B			UNITO
			Min.	Тур.	Max.	Min.	Тур.	Max.	UNITS
ISC	Short Circuit Current	H = 100 fc, 2850 K	4	5		4	5		μA
TC I _{SC}	ISC Temperature Coefficient	2850 K		.02	.08		.02	.08	%/°C
Voc	Open Circuit Voltage	H = 100 fc, 2850 K		420	3 0		420		mV
TC V _{OC}	V _{OC} Temperature Coefficient	2850 K		-2.0			-2.0		mV/°C
ID	Dark Current	H = 0, VR = 2.0 V			2000			100	pA
R _{SH}	Shunt Resistance	H = 0, V = 10 mV		.07			1.4		GΩ
TC R _{SH}	R _{SH} Temperature Coefficient	H = 0, V = 10 mV		-8.0			-8.0		%/°C
CJ	Junction Capacitance	H = 0, V = 0		1.0		2 1	1.0		nF
λ _{range}	Spectral Application Range		330		720	330		720	nm
λρ	Spectral Response - Peak			580			580		nm
VBR	Breakdown Voltage		2	40		2	40		V
θ1/2	Angular Resp 50% Resp. Pt.			±50			±50		Degrees
NEP	Noise Equivalent Power		1.1 x 10 ⁻¹³ (Typ.)			2.4 x 10 ⁻¹⁴ (Typ.)			W∕√Hz
D*	Specific Detectivity		2.2 x 10 ¹² (Typ.)			9.7 x 10 ¹² (Typ.)			cm_/Hz/W

VTB8440B photodiode



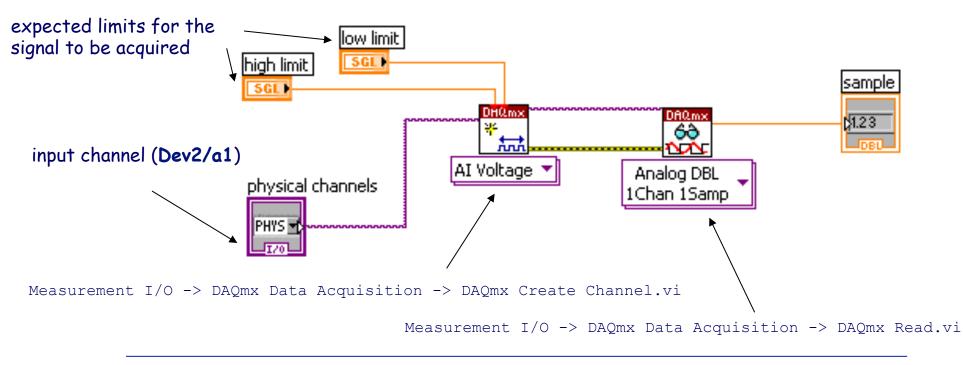


Front panel



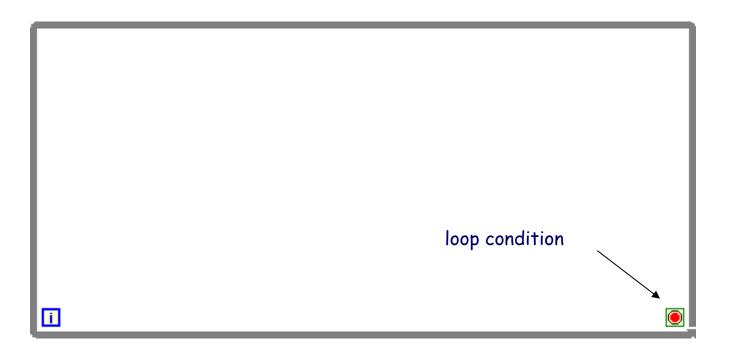
DAQmx Create Channel.vi and DAQmx Read.vi

- DAQmx Create Channel.vi provides the acquisition board with information about the type and range of the signals to be acquired and about the input channel
- DAQmx Read.vi samples the signal from the specified channel and yield the measured value



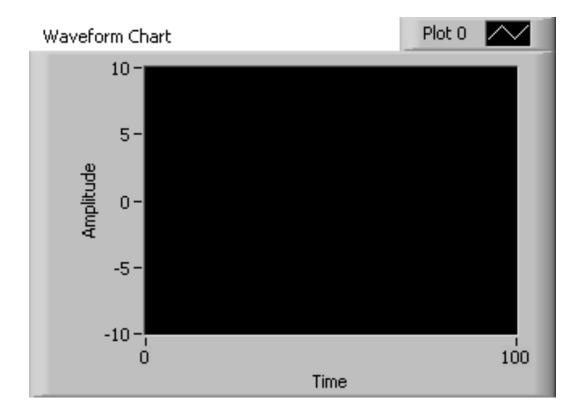
While loop

Needed for continuous acquisition of the signal coming from the conditioning circuit (you can find it in the Structures menu from the Functions palette) - a "stop" button should be included in the virtual instrument to stop the acquisition



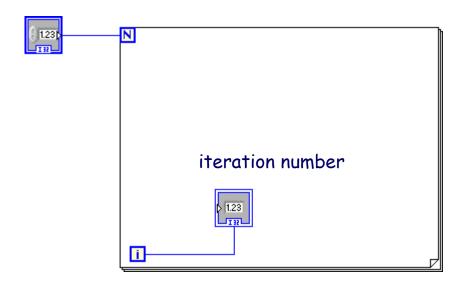
Waveform chart

You can use a waveform chart for a graphical representation of the acquired data ('Graph' menu of the Controls palette, from the front panel window) – the acquired sample can be directly fed to the waveform chart





We can use a for cycle to reduce the effects of zero average disturbances, therefore improving the measurement accuracy



Instead of representing (in the graph or in the numeric indicator) each individual acquired sample of the signal, we can represent the average value of N samples - the speed at which the measurement result is represented on the graph will decrease by a factor of N

number of cycles